- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line
- Approximations
- Plot Magnitude Response

+

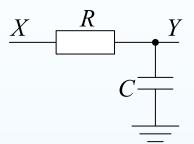
- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary

11: Frequency Responses

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If x(t) is a sine wave, then y(t) will also be a sine wave but with a different amplitude and phase shift. X is an input phasor and Y is the output phasor.



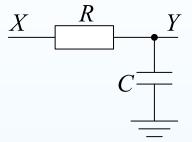
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The gain of the circuit is $\frac{Y}{X} = \frac{1/j\omega C}{R+1/j\omega C} = \frac{1}{j\omega RC+1}$

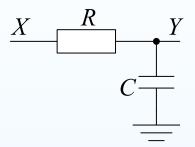
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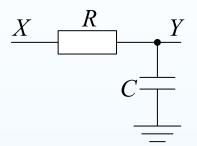


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$$\left|\frac{Y}{X}\right| = \frac{1}{|j\omega RC+1|} = \frac{1}{\sqrt{1+(\omega RC)^2}}$$

11: Frequency Responses

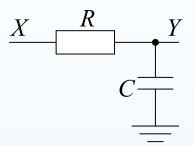
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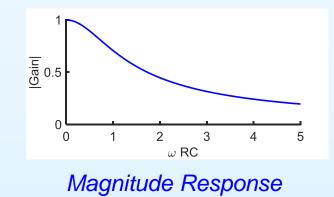
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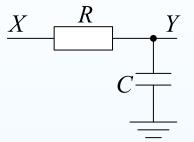
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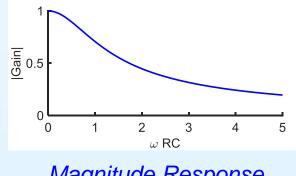
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Phase Shift: $\angle \left(\frac{Y}{X}\right) = -\angle (j\omega RC+1) = -\arctan\left(\frac{\omega R}{1}\right)$



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If x(t) is a sine wave, then y(t) will also be a sine wave but with a different amplitude and phase shift. X is an input phasor and Y is the output phasor.

 $X \xrightarrow{R} Y$

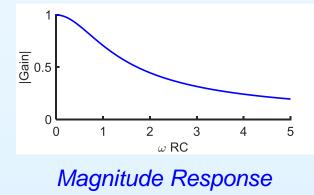
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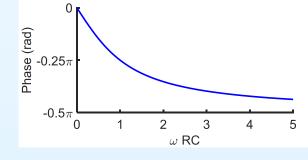
This is a complex function of ω so we plot separate graphs for:

Magnitude:
$$\left|\frac{Y}{X}\right| = \frac{1}{|j\omega RC+1|} = \frac{1}{\sqrt{1+(\omega RC)^2}}$$

Deco Shift: $\left\langle \begin{pmatrix} Y \\ Y \end{pmatrix} = -\left\langle (i_{X})PC + 1 \right\rangle = - \arctan\left(\frac{\omega R}{2}\right)$

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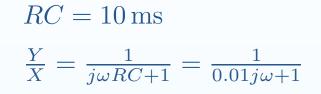


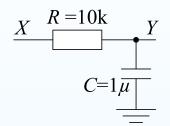
Phase Response

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$$\omega = 50 \Rightarrow \frac{Y}{X} = 0.89\angle -27^{\circ}$$

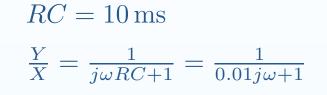
$$\omega = 100 \Rightarrow \frac{Y}{X} = 0.71\angle -45^{\circ}$$

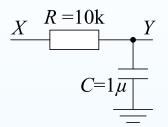
$$\omega = 300 \Rightarrow \frac{Y}{X} = 0.32\angle -72^{\circ}$$

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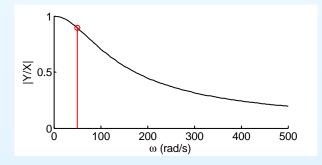


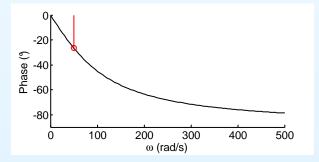


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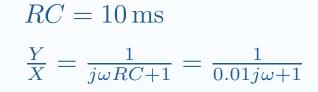


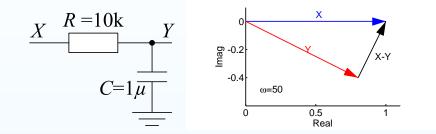


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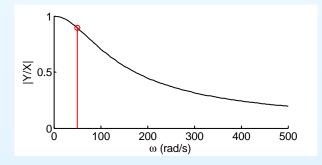
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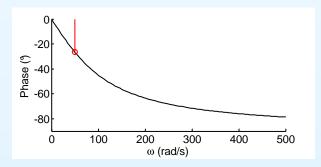




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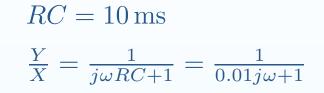




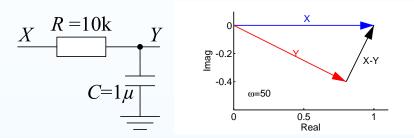
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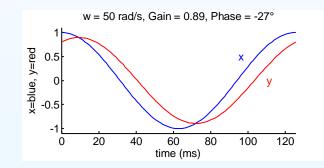
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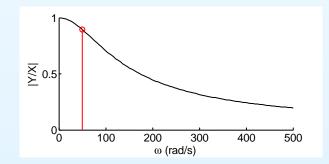
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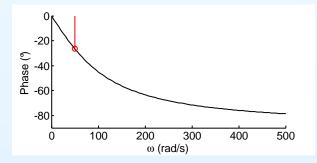


$$\omega = 50 \Rightarrow \frac{Y}{X} = 0.89\angle - 27^{\circ}$$
$$\omega = 100 \Rightarrow \frac{Y}{X} = 0.71\angle - 45^{\circ}$$
$$\omega = 300 \Rightarrow \frac{Y}{X} = 0.32\angle - 72^{\circ}$$





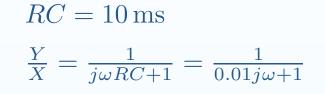


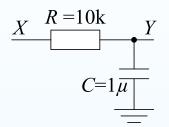


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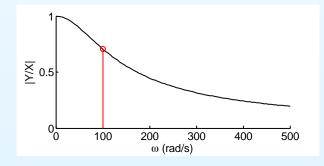


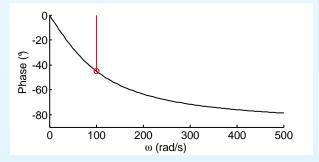


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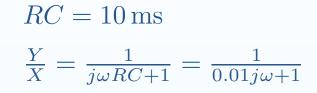


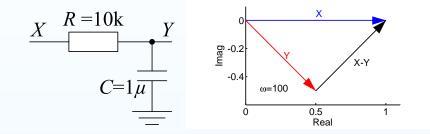


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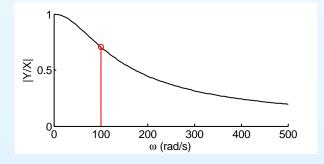


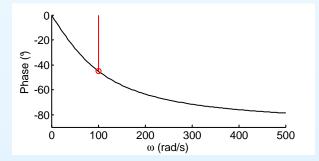


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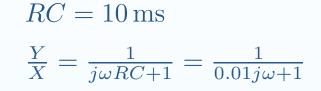


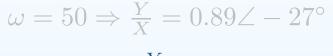


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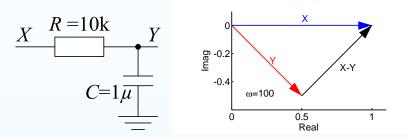
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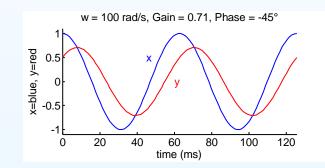
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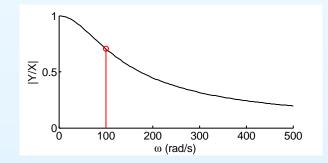


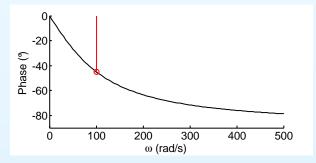


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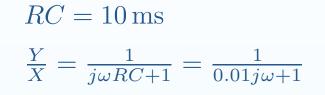


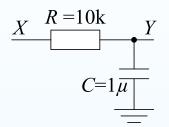


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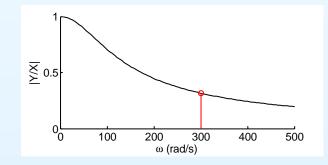


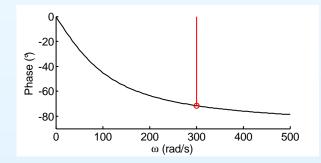


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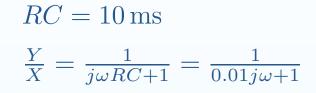


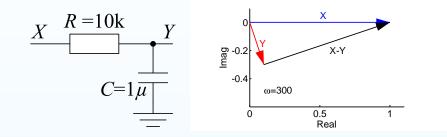


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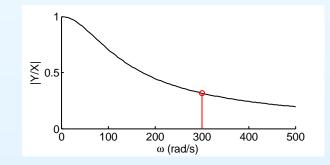


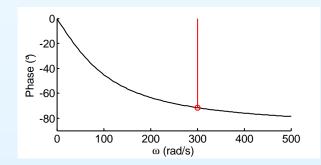


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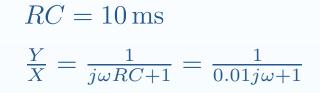


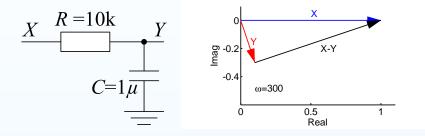


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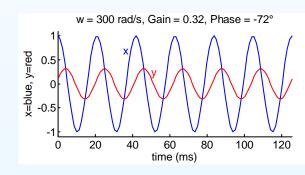


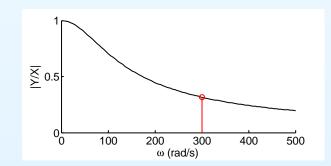


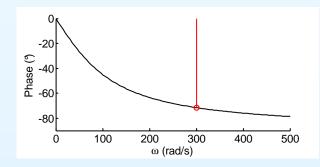
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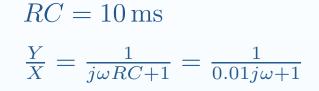


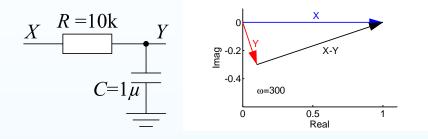
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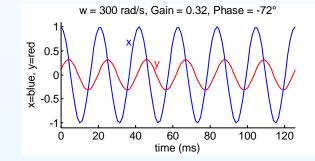


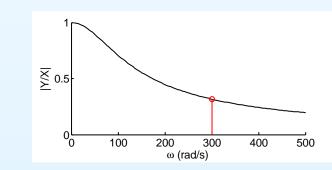


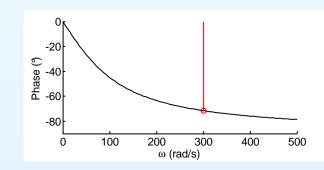
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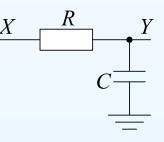


The output, y(t), *lags* the input, x(t), by up to 90° .

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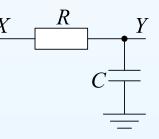
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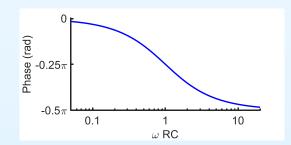
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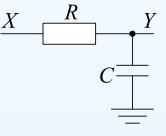


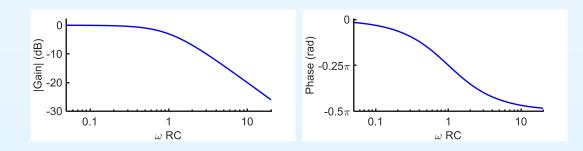


11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line Approximations
- Plot Magnitude Response
- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary

We usually use logarithmic axes for frequency and gain (but not phase) because % differences are more significant than absolute differences. E.g. 5 kHz versus 5.005 kHz is less significant than 10 Hz versus 15 Hz even though both differences equal 5 Hz.





E1.1 Analysis of Circuits (2018-10340)

Frequency Responses: 11 - 4 / 12

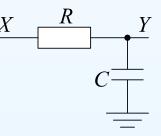
11: Frequency Responses

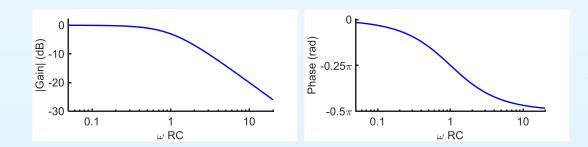
- Frequency Response
- Sine Wave Response
- Logarithmic axes
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- Straight Line Approximations
- Plot Magnitude Response
- Low and High Frequency Asymptotes
- Phase Approximation

+

- Plot Phase Response +
- RCR Circuit
- Summary

We usually use logarithmic axes for frequency and gain (but not phase) because % differences are more significant than absolute differences. E.g. 5 kHz versus 5.005 kHz is less significant than 10 Hz versus 15 Hz even though both differences equal 5 Hz.





Note that 0 does not exist on a log axis and so the starting point of the axis is arbitrary.

E1.1 Analysis of Circuits (2018-10340)

11: Frequency Responses

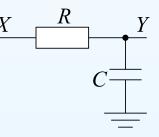
- Frequency Response
- Sine Wave Response
- Logarithmic axes
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- Plot Phase Response

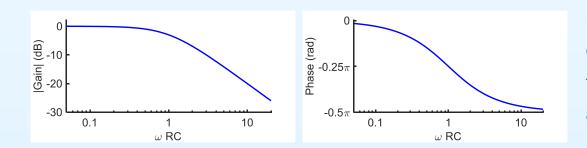
+

- RCR Circuit
- Summary

We usually use logarithmic axes for frequency and gain (but not phase) because % differences are more significant than absolute differences. E.g. 5 kHz versus 5.005 kHz is less significant than 10 Hz versus 15 Hz even though both differences equal 5 Hz.

Logarithmic voltage ratios are specified in *decibels* (dB) = $20 \log_{10} \frac{|V_2|}{|V_1|}$.





11: Frequency Responses

- Frequency Response
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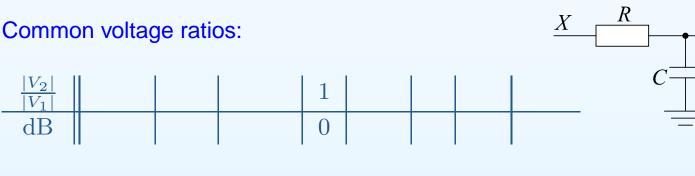
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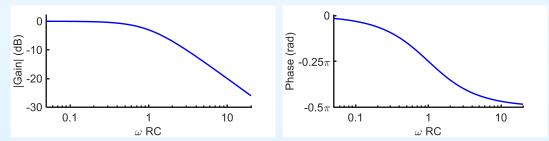
dB

- RCR Circuit
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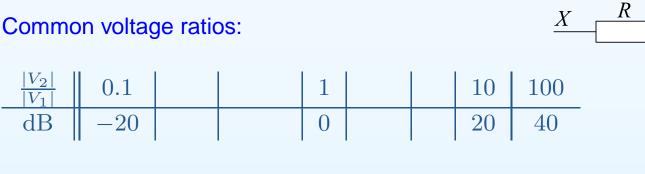
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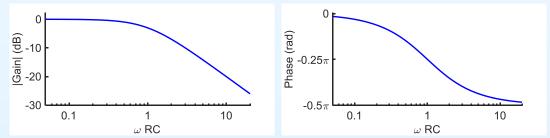
dB

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Logarithmic voltage ratios are specified in *decibels* (dB) = $20 \log_{10} \frac{|V_2|}{|V_1|}$.

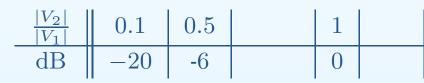
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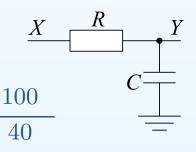
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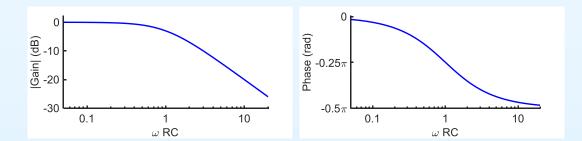
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11: Frequency Responses

- Frequency Response
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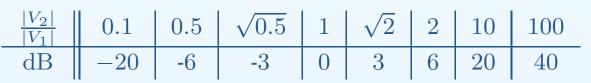
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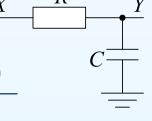
- RCR Circuit
- Summary

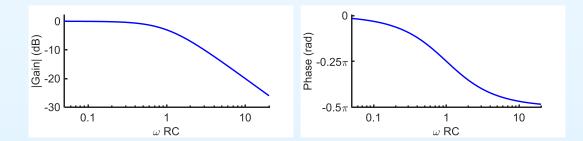
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11: Frequency Responses

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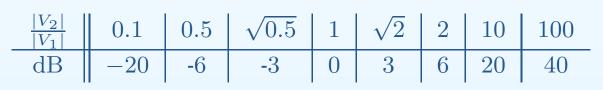
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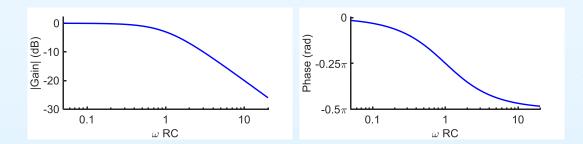
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Logarithmic voltage ratios are specified in *decibels* (dB) = $20 \log_{10} \frac{|V_2|}{|V_1|}$.







Note that 0 does not exist on a log axis and so the starting point of the axis is arbitrary.

Note: $P \propto V^2 \Rightarrow$ decibel <u>power</u> ratios are given by $10 \log_{10} \frac{P_2}{P_1}$

Frequency Responses: 11 – 4 / 12

11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line Approximations
- Plot Magnitude Response

+

- Low and High Frequency Asymptotes
- Phase Approximation +
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Suppose we plot the magnitude and phase of $H = c \left(j \omega \right)^r$

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Magnitude (log-log graph):

$$|H| = c\omega^r \Rightarrow \log|H| = \log|c| + r\log\omega$$

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- Frequency Response
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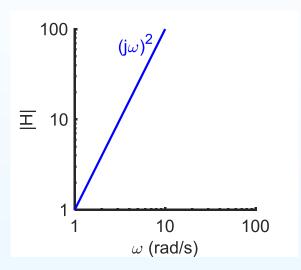
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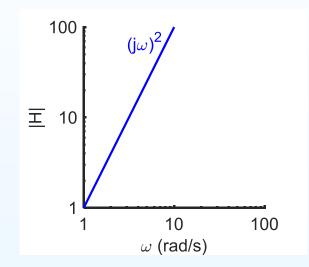
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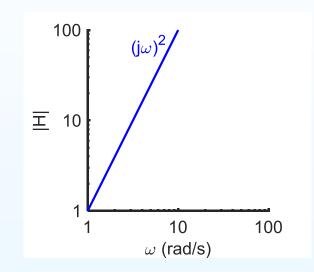
11: Frequency Responses

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Phase (log-lin graph):

 $\angle H = \angle j^r + \angle c = r \times \frac{\pi}{2}$ (+ π if c < 0)

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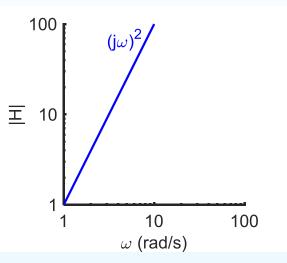
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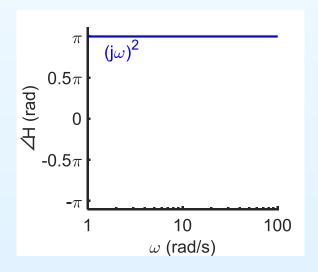
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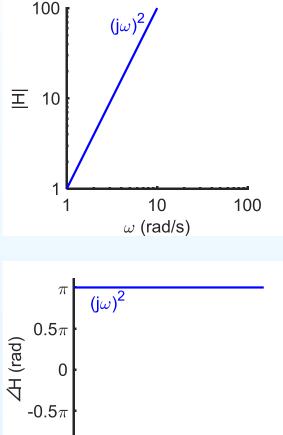
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Suppose we plot the magnitude and phase of $H = c (j\omega)^r$

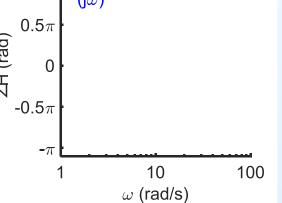
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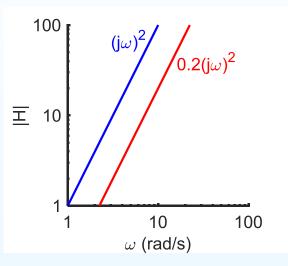
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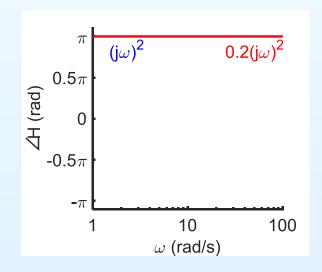
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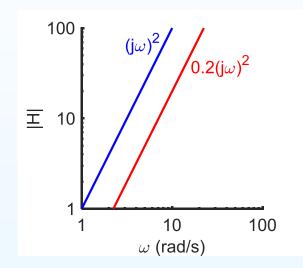
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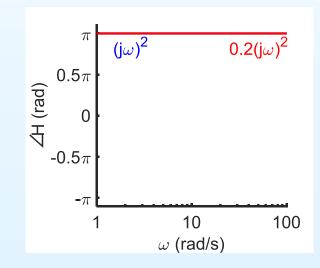
Phase (log-lin graph):

The phase is constant $\forall \omega$.

 $|H| = c\omega^r \Rightarrow \log |H| = \log |c| + r \log \omega$ This is a straight line with a slope of r. c only affects the line's vertical position.

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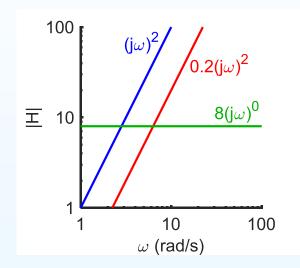
- Plot Phase Response +
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Plot Magnitude Response

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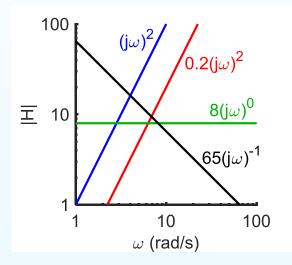
Magnitude (log-log graph):

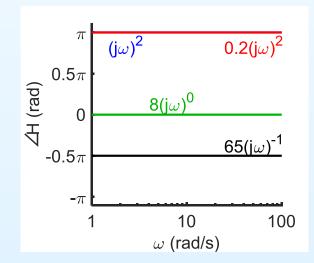
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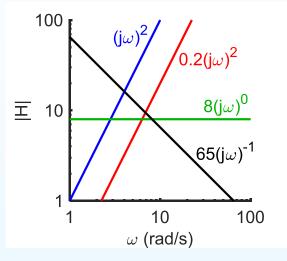
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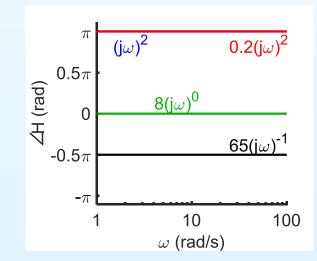
- Plot Phase Response +
- RCR Circuit
- Summary

Suppose we plot the magnitude and phase of $H = c \left(j \omega \right)^r$

Magnitude (log-log graph):

 $|H| = c\omega^r \Rightarrow \log |H| = \log |c| + r \log \omega$ This is a straight line with a slope of r. c only affects the line's vertical position.





Phase (log-lin graph):

 $\angle H = \angle j^r + \angle c = r \times \frac{\pi}{2} \ (+\pi \text{ if } c < 0)$

The phase is constant $\forall \omega$.

If c > 0, phase = $90^{\circ} \times$ magnitude slope.

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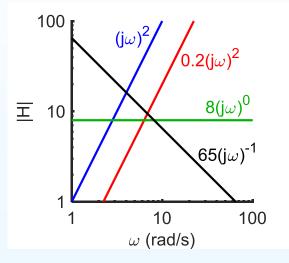
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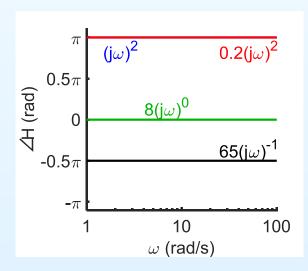
If |H| is measured in decibels, a slope of r is called $6r \, d\text{B/octave}$ or $20r \, d\text{B/decade}$.



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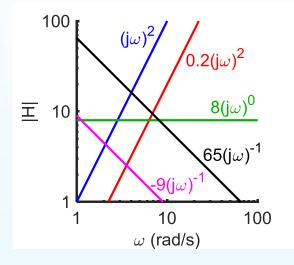
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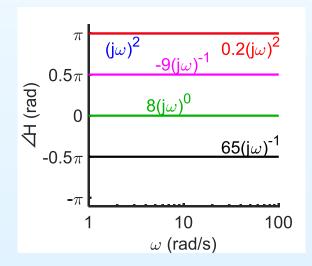
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If c > 0, phase = $90^{\circ} \times$ magnitude slope. Negative c adds $\pm 180^{\circ}$ to the phase.





11: Frequency Responses

- Frequency Response
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Suppose we plot the magnitude and phase of $H = c \left(j \omega \right)^r$

Magnitude (log-log graph):

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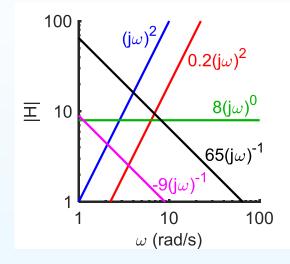
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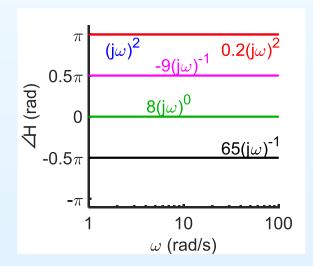
Phase (log-lin graph):

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If c > 0, phase = $90^{\circ} \times$ magnitude slope. Negative c adds $\pm 180^{\circ}$ to the phase.

Note: Phase angles are modulo 360° , i.e. $+180^{\circ} \equiv -180^{\circ}$ and $450^{\circ} \equiv 90^{\circ}$.





affects the line's vertical position.

11: Frequency Responses

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- RCR Circuit
- Summary

 $H = c (j\omega)^r$ has a straight-line magnitude graph and a constant phase.

Magnitude (log-log graph):

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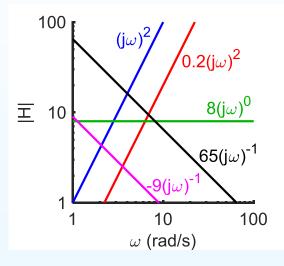
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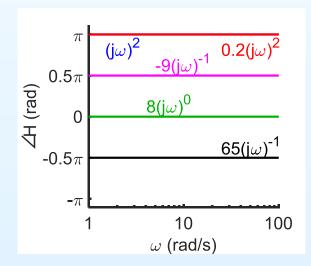
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 $\angle H = \angle j^r + \angle c = r \times \frac{\pi}{2}$ (+ π if c < 0) The phase is constant $\forall \omega$.

If c > 0, phase = $90^{\circ} \times$ magnitude slope. Negative c adds $\pm 180^{\circ}$ to the phase.

Note: Phase angles are modulo 360° , i.e. $+180^{\circ} \equiv -180^{\circ}$ and $450^{\circ} \equiv 90^{\circ}$.





11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers

Straight Line

- Approximations
- Plot Magnitude Response

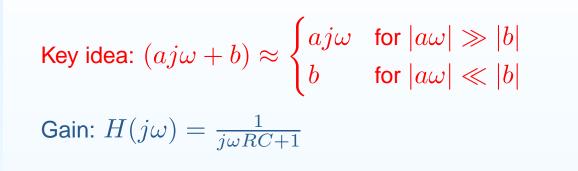
- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary

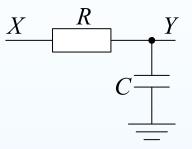
Key idea: $(aj\omega + b) \approx \begin{cases} aj\omega & \text{for } |a\omega| \gg |b| \\ b & \text{for } |a\omega| \ll |b| \end{cases}$

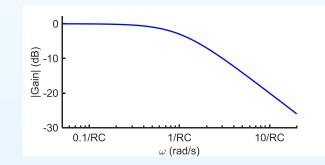
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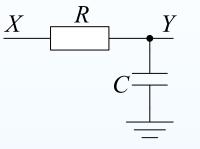
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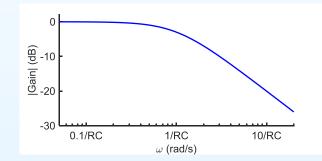
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Gain: $H(j\omega) = \frac{1}{j\omega RC + 1}$



Low frequencies (
$$\omega \ll \frac{1}{RC}$$
): $H(j\omega) \approx 1$



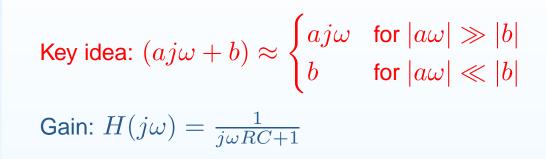
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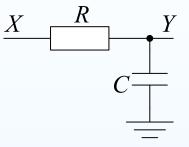
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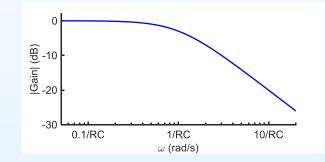
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Low frequencies ($\omega \ll \frac{1}{RC}$): $H(j\omega) \approx 1$ High frequencies ($\omega \gg \frac{1}{RC}$): $H(j\omega) \approx \frac{1}{j\omega RC}$



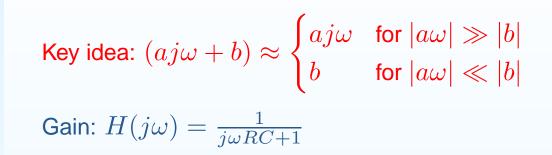
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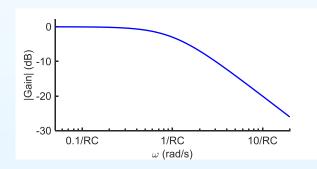
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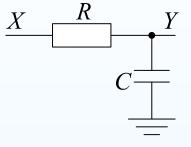
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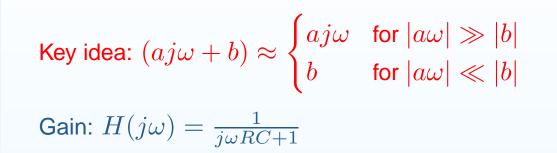
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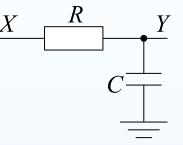
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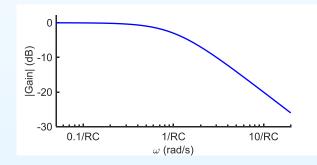
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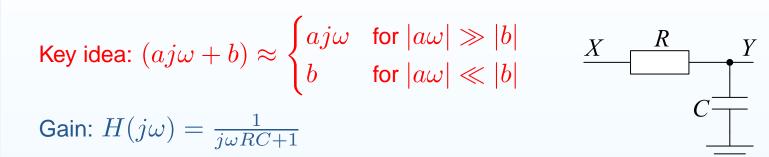
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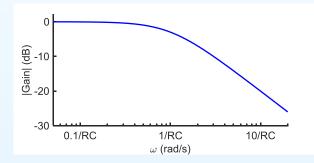
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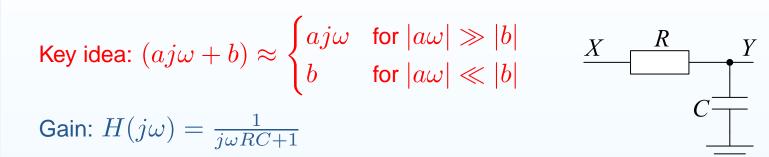
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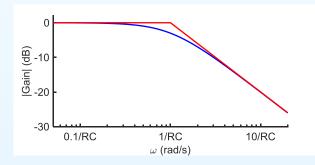
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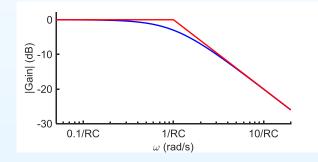
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 $\begin{array}{ll} \text{Key idea: } (aj\omega + b) \approx \begin{cases} aj\omega & \text{for } |a\omega| \gg |b| \\ b & \text{for } |a\omega| \ll |b| \end{cases} & \underbrace{X \quad R} \\ \hline \\ \text{Gain: } H(j\omega) = \frac{1}{j\omega RC + 1} & C^{-1} \end{array}$

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Approximate the magnitude response as two straight lines intersecting at the <u>corner frequency</u>, $\omega_c = \frac{1}{RC}$.



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(a) the gradient changes by -1 (= -6 dB/octave = -20 dB/decade).

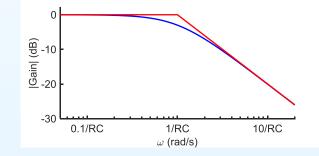
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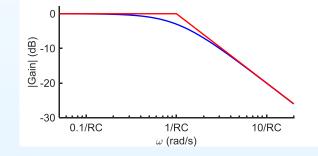
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A linear factor $(aj\omega + b)$ has a corner frequency of $\omega_c = \left|\frac{b}{a}\right|$.

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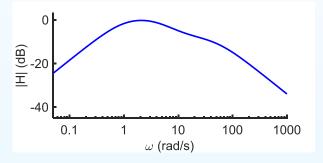
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The gain of a linear circuit is always a *rational polynomial* in $j\omega$ and is called the *transfer function* of the circuit. For example:

 $H(j\omega) = \frac{60(j\omega)^2 + 720(j\omega)}{3(j\omega)^3 + 165(j\omega)^2 + 762(j\omega) + 600}$



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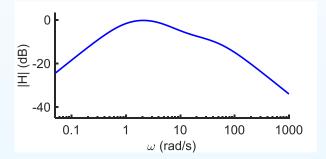
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Step 1: Factorize the polynomials



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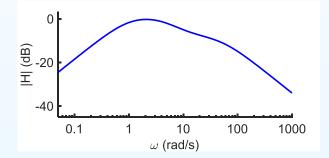
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Step 1: Factorize the polynomials Step 2: Sort corner freqs: 1, 4, 12, 50



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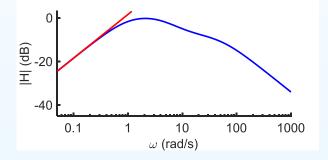
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$$|H| \approx \frac{20\omega \times 12}{1 \times 4 \times 50} = 1.2\omega^1.$$



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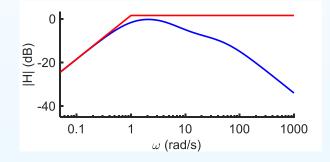
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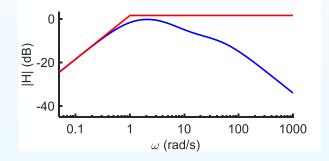
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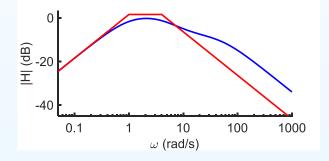
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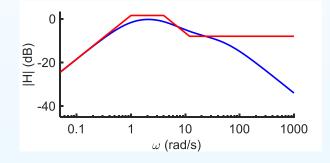
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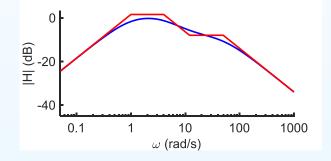
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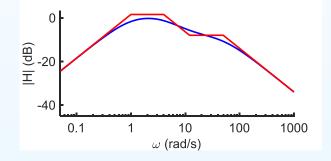
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Step 1: Factorize the polynomials Step 2: Sort corner freqs: 1, 4, 12, 50 Step 3: For $\omega < 1$ all linear factors equal their constant terms: $|H| \approx \frac{20\omega \times 12}{1\times 4\times 50} = 1.2\omega^{1}.$



Step 4: For $1 < \omega < 4$, the factor $(j\omega + 1) \approx j\omega$ so $|H| \approx \frac{20\omega \times 12}{\omega \times 4 \times 50} = 1.2\omega^0 = +1.58 \,\mathrm{dB}.$ Step 5: For $4 < \omega < 12$, $|H| \approx \frac{20\omega \times 12}{\omega \times \omega \times 50} = 4.8\omega^{-1}.$ Step 6: For $12 < \omega < 50$, $|H| \approx \frac{20\omega \times \omega}{\omega \times \omega \times 50} = 0.4\omega^0 = -7.96 \,\mathrm{dB}.$ Step 7: For $\omega > 50$, $|H| \approx \frac{20\omega \times \omega}{\omega \times \omega \times \omega} = 20\omega^{-1}.$

At each corner frequency, the graph is continuous but its gradient changes abruptly by +1 (numerator factor) or -1 (denominator factor).

11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line
- Approximations
- Plot Magnitude Response

+

- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary

You can find the low and high frequency asymptotes without factorizing: $H(j\omega) = \frac{60(j\omega)^2 + 720(j\omega)}{3(j\omega)^3 + 165(j\omega)^2 + 762(j\omega) + 600} = \frac{20j\omega(j\omega+12)}{(j\omega+1)(j\omega+4)(j\omega+50)}$

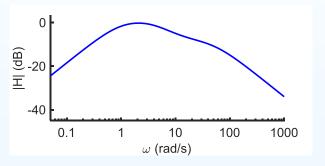
11: Frequency Responses

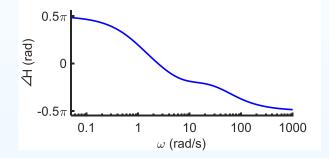
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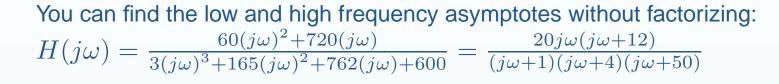
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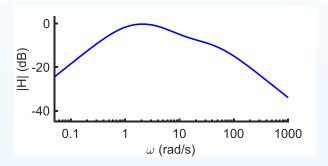
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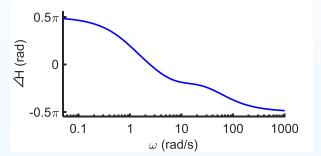
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Low Frequency Asymptote:

11: Frequency Responses

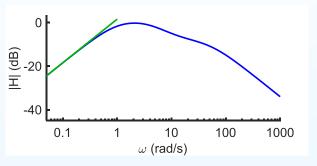
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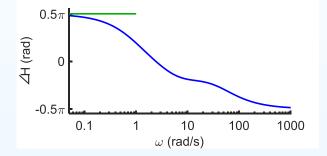
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Low Frequency Asymptote: From factors: $H_{\rm LF}(j\omega) = \frac{20j\omega(12)}{(1)(4)(50)} = 1.2j\omega$

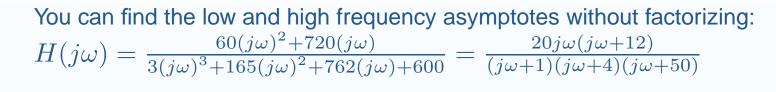
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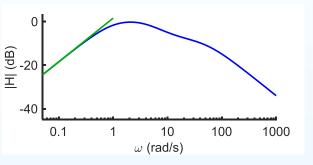
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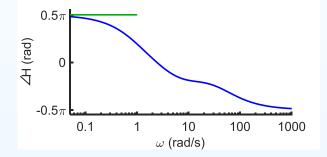
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Low Frequency Asymptote: From factors: $H_{\rm LF}(j\omega) = \frac{20j\omega(12)}{(1)(4)(50)} = 1.2j\omega$ Lowest power of $j\omega$ on top and bottom: $H(j\omega) \simeq \frac{720(j\omega)}{600} = 1.2j\omega$

Low and High Frequency Asymptotes

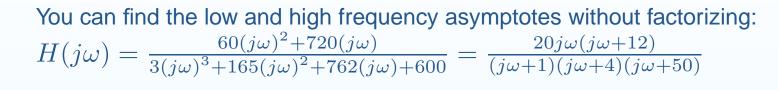
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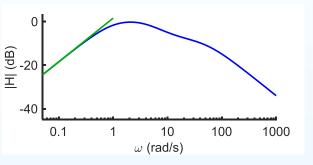
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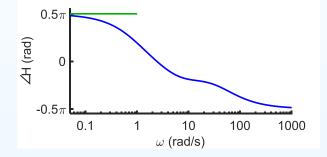
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Low and High Frequency Asymptotes

11: Frequency Responses

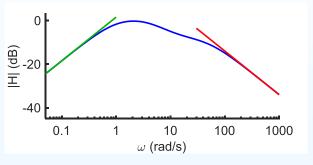
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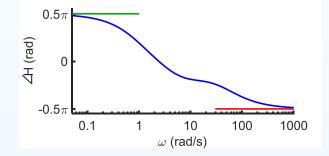
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Low Frequency Asymptote: From factors: $H_{\text{LF}}(j\omega) = \frac{20j\omega(12)}{(1)(4)(50)} = 1.2j\omega$ Lowest power of $j\omega$ on top and bottom: $H(j\omega) \simeq \frac{720(j\omega)}{600} = 1.2j\omega$ High Frequency Asymptote:

From factors: $H_{\rm HF}(j\omega) = \frac{20j\omega(j\omega)}{(j\omega)(j\omega)(j\omega)} = 20 (j\omega)^{-1}$

Low and High Frequency Asymptotes

11: Frequency Responses

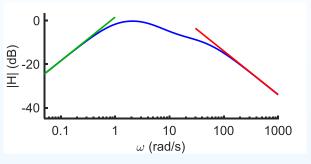
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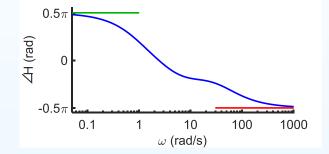
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You can find the low and high frequency asymptotes without factorizing: $H(j\omega) = \frac{60(j\omega)^2 + 720(j\omega)}{3(j\omega)^3 + 165(j\omega)^2 + 762(j\omega) + 600} = \frac{20j\omega(j\omega+12)}{(j\omega+1)(j\omega+4)(j\omega+50)}$



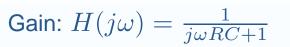


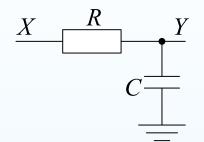
Low Frequency Asymptote: From factors: $H_{\rm LF}(j\omega) = \frac{20j\omega(12)}{(1)(4)(50)} = 1.2j\omega$ Lowest power of $j\omega$ on top and bottom: $H(j\omega) \simeq \frac{720(j\omega)}{600} = 1.2j\omega$ High Frequency Asymptote: From factors: $H_{\rm HF}(j\omega) = \frac{20j\omega(j\omega)}{(j\omega)(j\omega)(j\omega)} = 20 (j\omega)^{-1}$ Highest power of $j\omega$ on top and bottom: $H(j\omega) \simeq \frac{60(j\omega)^2}{3(j\omega)^3} = 20 (j\omega)^{-1}$

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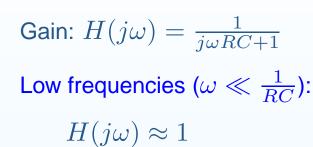


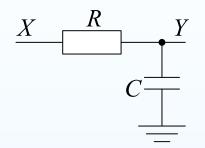


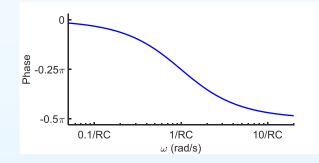
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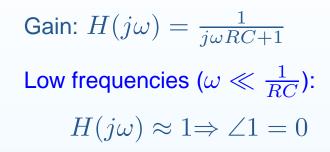


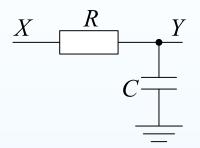
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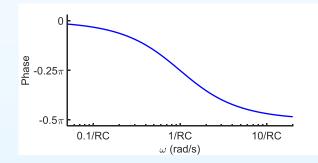
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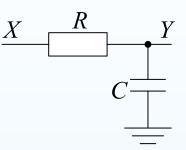




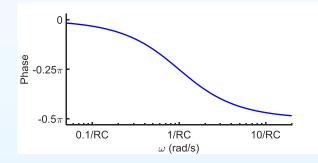
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Gain: $H(j\omega) = \frac{1}{j\omega RC + 1}$ Low frequencies ($\omega \ll \frac{1}{RC}$): $H(j\omega) \approx 1 \Rightarrow \angle 1 = 0$



High frequencies ($\omega\gg\frac{1}{RC}$): $H(j\omega)\approx\frac{1}{j\omega RC}$

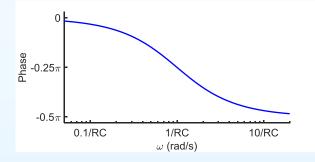


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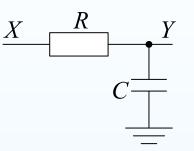


E1.1 Analysis of Circuits (2018-10340)

11: Frequency Responses

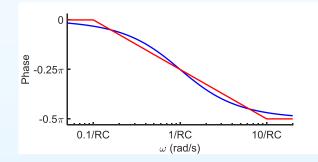
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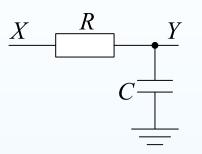
Approximate the phase response as three straight lines.



11: Frequency Responses

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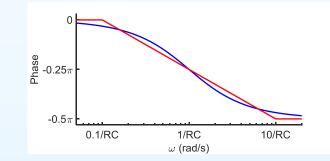
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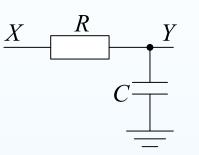
By chance, they intersect close to $0.1\omega_c$ and $10\omega_c$ where $\omega_c = \frac{1}{RC}$.



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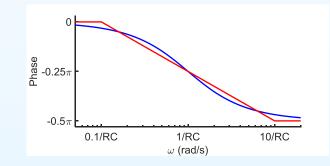
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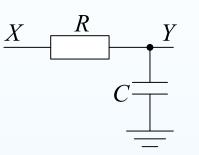


Between $0.1\omega_c$ and $10\omega_c$ the phase changes by $-\frac{\pi}{2}$ over two decades. This gives a gradient = $-\frac{\pi}{4}$ radians/decade.

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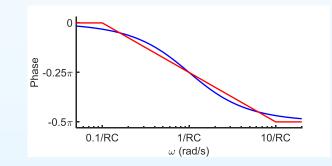
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 $(aj\omega + b)$ in denominator $\Rightarrow \Delta \text{gradient} = \mp \frac{\pi}{4}/\text{decade at } \omega = 10^{\mp 1} \left| \frac{b}{a} \right|.$

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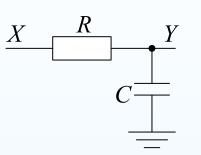
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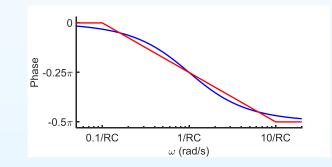
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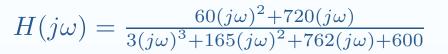
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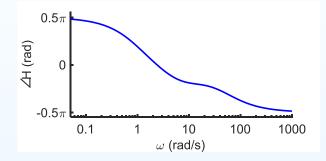
The sign of Δ gradient is reversed for (a) numerator factors and (b) $\frac{b}{a} < 0$.

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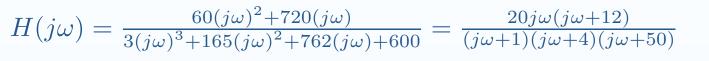


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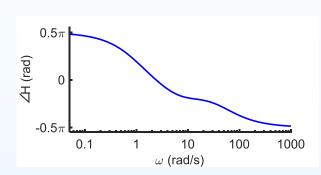
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Step 1: Factorize the polynomials Step 2: List corner freqs: $\pm =$ num/den $\omega_c = \{1^-, 4^-, 12^+, 50^-\}$



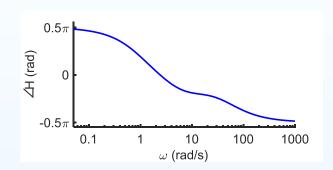
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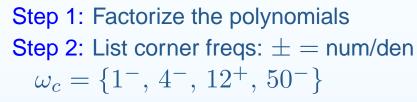
Step 3: Gradient changes at $10^{\pm 1}\omega_c$. Sign depends on num/den and sgn $(\frac{b}{a})$: $.1^-, 10^+; .4^-, 40^+; 1.2^+, 120^-; 5^-, 500^+$



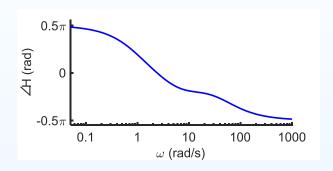
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 $H(j\omega) = \frac{60(j\omega)^2 + 720(j\omega)}{3(j\omega)^3 + 165(j\omega)^2 + 762(j\omega) + 600} = \frac{20j\omega(j\omega+12)}{(j\omega+1)(j\omega+4)(j\omega+50)}$



Step 3: Gradient changes at $10^{\pm 1}\omega_c$. Sign depends on num/den and sgn $(\frac{b}{a})$: .1⁻, 10⁺; .4⁻, 40⁺; 1.2⁺, 120⁻; 5⁻, 500⁺

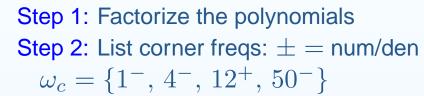


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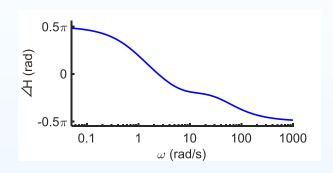
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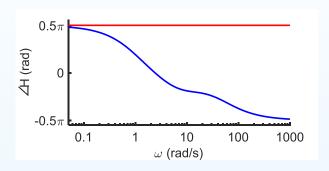
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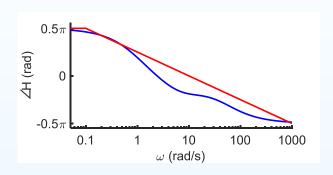
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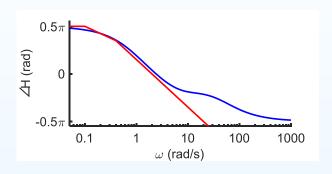
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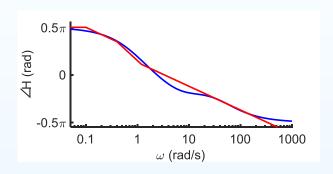
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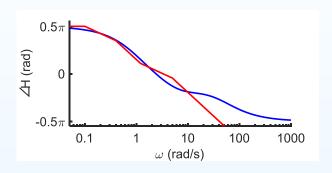
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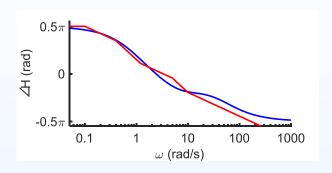
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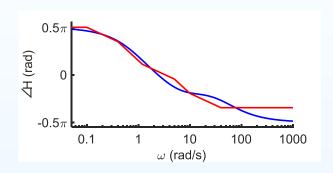
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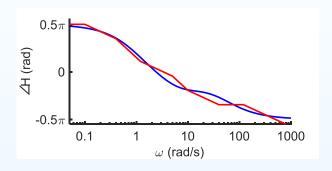
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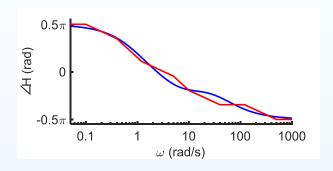
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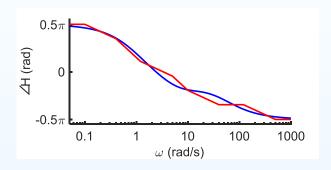
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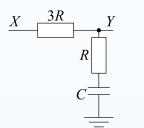


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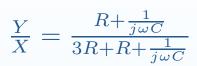
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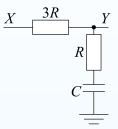


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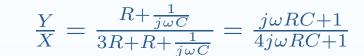


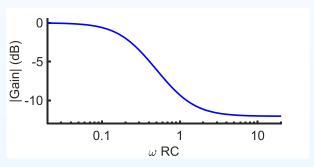


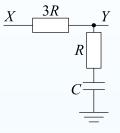
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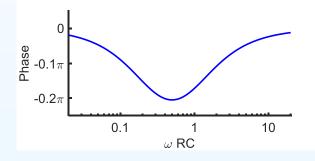
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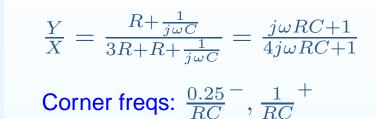


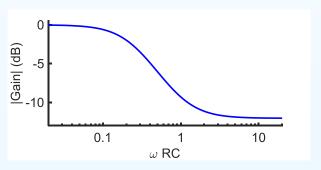


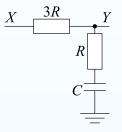
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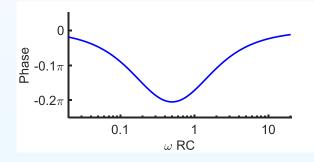
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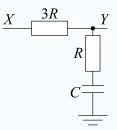
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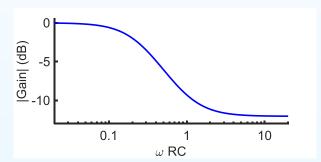
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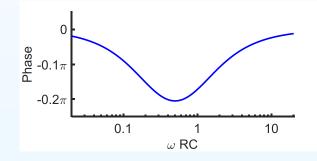
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 $\frac{Y}{X} = \frac{R + \frac{1}{j\omega C}}{3R + R + \frac{1}{j\omega C}} = \frac{j\omega RC + 1}{4j\omega RC + 1}$

Corner freqs:
$$\frac{0.25}{RC}^{-}, \frac{1}{RC}^{+}$$
 LF Asymptote: $H(j\omega) = 1$







11: Frequency Responses

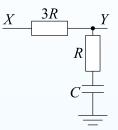
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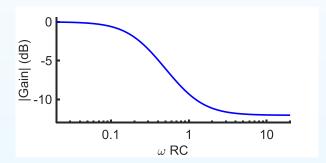
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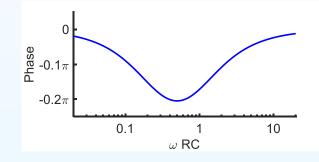
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Magnitude Response:

0

-5

-10

Gain (dB)

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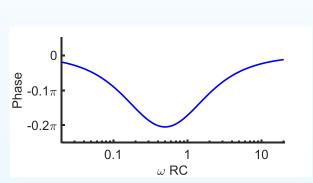
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Magnitude Response:

1

 ωRC

0.1

Gradient Changes: $-20 \, \mathrm{dB}/\mathrm{dec}$ at $\omega = \frac{0.25}{RC}$ and +20 at $\omega = \frac{1}{RC}$

E1.1 Analysis of Circuits (2018-10340)

0

-5

-10

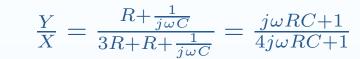
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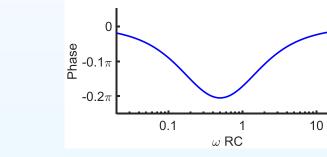
 $4j\omega RC^{-1}$

1

 ωRC

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Magnitude Response:

0.1

Gradient Changes: $-20 \, \mathrm{dB}/\mathrm{dec}$ at $\omega = \frac{0.25}{RC}$ and +20 at $\omega = \frac{1}{RC}$

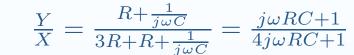
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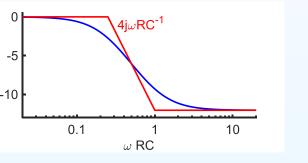
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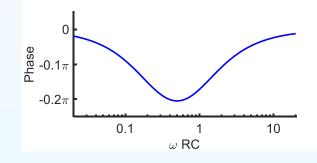
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 LF Asymptote: $H(j\omega) = 1$





Magnitude Response:

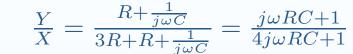
Gradient Changes: -20 dB/dec at $\omega = \frac{0.25}{RC}$ and $+20 \text{ at } \omega = \frac{1}{RC}$ Line equations: $H(j\omega) =$ (a) 1, (b) $\frac{1}{4j\omega RC}$, (c) $\frac{j\omega RC}{4j\omega RC} = 0.25$

11: Frequency Responses

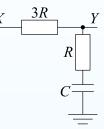
- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line
- Approximations
- Plot Magnitude Response

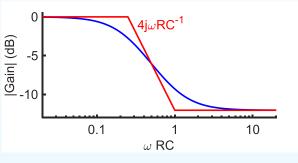
+

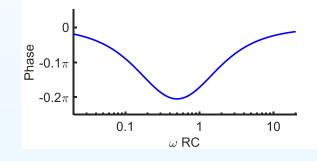
- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary



Corner freqs:
$$\frac{0.25}{RC}^{-}, \frac{1}{RC}^{+}$$
 LF Asymptote: $H(j\omega) = 1$







Magnitude Response:

Gradient Changes: -20 dB/dec at $\omega = \frac{0.25}{RC}$ and $+20 \text{ at } \omega = \frac{1}{RC}$ Line equations: $H(j\omega) =$ (a) 1, (b) $\frac{1}{4j\omega RC}$, (c) $\frac{j\omega RC}{4j\omega RC} = 0.25$

Phase Response:

11: Frequency Responses

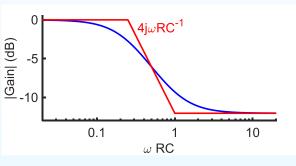
- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line
- Approximations
- Plot Magnitude Response

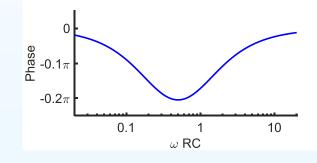
+

- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary

 $\frac{Y}{X} = \frac{R + \frac{1}{j\omega C}}{3R + R + \frac{1}{j\omega C}} = \frac{j\omega RC + 1}{4j\omega RC + 1}$

Corner freqs:
$$\frac{0.25}{RC}^{-}, \frac{1}{RC}^{+}$$
 LF Asymptote: $H(j\omega) = 1$





Magnitude Response:

Gradient Changes: -20 dB/dec at $\omega = \frac{0.25}{RC}$ and $+20 \text{ at } \omega = \frac{1}{RC}$ Line equations: $H(j\omega) =$ (a) 1, (b) $\frac{1}{4j\omega RC}$, (c) $\frac{j\omega RC}{4j\omega RC} = 0.25$

Phase Response:

LF asymptote: $\phi = \angle 1 = 0$

n

-5

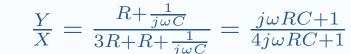
Gain (dB)

11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line
- Approximations
- Plot Magnitude Response

+

- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary



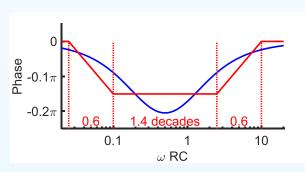
 $4j\omega RC^{-1}$

1

 ωRC

Corner freqs:
$$\frac{0.25}{RC}^{-}, \frac{1}{RC}^{+}$$
 LF Asymptote: $H(j\omega) = 1$

10



Magnitude Response:

0.1

Gradient Changes: -20 dB/dec at $\omega = \frac{0.25}{RC}$ and $+20 \text{ at } \omega = \frac{1}{RC}$ Line equations: $H(j\omega) =$ (a) 1, (b) $\frac{1}{4j\omega RC}$, (c) $\frac{j\omega RC}{4j\omega RC} = 0.25$

Phase Response:

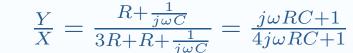
LF asymptote: $\phi = \angle 1 = 0$ Gradient changes of $\pm \frac{\pi}{4}$ /decade at: $\omega = \frac{0.025}{RC}^{-}, \frac{0.1}{RC}^{+}, \frac{2.5}{RC}^{+}, \frac{10}{RC}^{-}$.

11: Frequency Responses

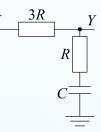
- Frequency Response
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- Straight Line
- Approximations
- Plot Magnitude Response

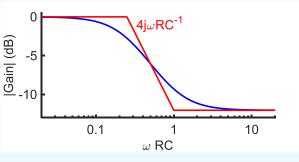
+

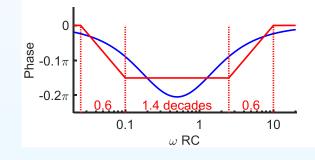
- Low and High Frequency Asymptotes
- Phase Approximation
- Plot Phase Response +
- RCR Circuit
- Summary



Corner freqs:
$$\frac{0.25}{RC}^{-}, \frac{1}{RC}^{+}$$
 LF Asymptote: $H(j\omega) = 1$







Magnitude Response:

Gradient Changes: -20 dB/dec at $\omega = \frac{0.25}{RC}$ and $+20 \text{ at } \omega = \frac{1}{RC}$ Line equations: $H(j\omega) =$ (a) 1, (b) $\frac{1}{4j\omega RC}$, (c) $\frac{j\omega RC}{4j\omega RC} = 0.25$

Phase Response:

LF asymptote: $\phi = \angle 1 = 0$ Gradient changes of $\pm \frac{\pi}{4}$ /decade at: $\omega = \frac{0.025}{RC}^{-}, \frac{0.1}{RC}^{+}, \frac{2.5}{RC}^{+}, \frac{10}{RC}^{-}$. At $\omega = \frac{0.1}{RC}, \phi = 0 - \frac{\pi}{4} \log_{10} \frac{0.1}{0.025} = -\frac{\pi}{4} \times 0.602 = -0.15\pi$

11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
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+

- Low and High Frequency Asymptotes
- Phase Approximation +
- Plot Phase Response +
- RCR Circuit
- Summary

• Frequency response: magnitude and phase of $\frac{Y}{X}$ as a function of ω

• Only applies to sine waves

11: Frequency Responses

- Frequency Response
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- Low and High Frequency Asymptotes
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- Frequency response: magnitude and phase of $\frac{Y}{X}$ as a function of ω
 - Only applies to sine waves
 - Use log axes for frequency and gain but linear for phase

▷ **Decibels** =
$$20 \log_{10} \frac{V_2}{V_1} = 10 \log_{10} \frac{P_2}{P_1}$$

11: Frequency Responses

- Frequency Response
- Sine Wave Response
- Logarithmic axes
- Logs of Powers
- Straight Line
- Approximations
- Plot Magnitude Response
- Low and High Frequency Asymptotes
- Phase Approximation
- Plot Phase Response
- RCR Circuit
- Summary

Frequency response: magnitude and phase of ^Y/_X as a function of ω
 Only applies to sine waves

• Use log axes for frequency and gain but linear for phase

▷ **Decibels =**
$$20 \log_{10} \frac{V_2}{V_1} = 10 \log_{10} \frac{P_2}{P_1}$$

- Linear factor $(aj\omega + b)$ gives corner frequency at $\omega = \left|\frac{b}{a}\right|$.
 - Magnitude plot gradient changes by $\pm 20 \, dB/decade @ \omega = \left| \frac{b}{a} \right|.$

11: Frequency Responses

- Frequency Response
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 - Phase gradient changes in two places by:
 - $\triangleright \pm \frac{\pi}{4} \text{ rad/decade } @ \omega = 0.1 \times \left| \frac{b}{a} \right|$
 - $\triangleright \quad \mp \frac{\pi}{4} \text{ rad/decade } @ \omega = 10 \times \left| \frac{b}{a} \right|$

11: Frequency Responses

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- LF/HF asymptotes: keep only the terms with the lowest/highest power of $j\omega$ in numerator and denominator polynomials

11: Frequency Responses

- Frequency Response
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For further details see Hayt Ch 16 or Irwin Ch 12.